

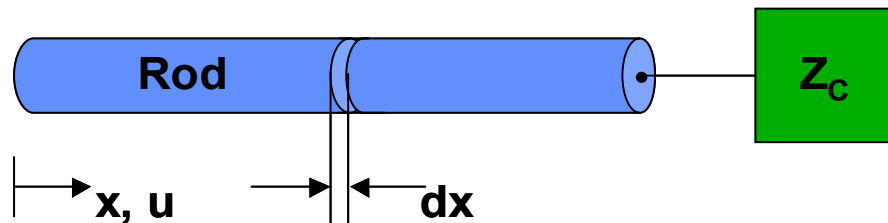
# Model <1> for Magnetostrictive Rod

*In this notebook the impedance of an ideal solenoid around a magnetostrictive rod is derived. Also derived is the amplitude of oscillations of the rod, dependant on the applied coil voltage. The model is only valid for steady-state oscillations, no electromechanical losses, and an ideal magnetic flux return path (that means the magnetic permeability  $\mu$  of the surrounding medium has to be significantly higher than the permeability of the rod itself).*

*Additionally this model simulates the electric circuit for a solenoid connected to a function generator, a voltage-source with an internal resistance.*

## ■ Initialization for Mathematica

## ■ General equations



## ■ Newton's second law

Newton's second law, reduced to a one-dimensional problem

```
In[19] := newton2 =  $\partial_x \sigma[x, t] == \rho R * \partial_{t,t} u[x, t]$ 
```

```
Out[19] =  $\sigma^{(1,0)}[x, t] == \rho R u^{(0,2)}[x, t]$ 
```

## ■ Strain $\epsilon$

Strain is the first partial derivative of displacement, with respect to the position:

```
In[20] :=  $\epsilon[\mathbf{x}_-, t_-] = \partial_{\mathbf{x}} u[\mathbf{x}, t]$ 
```

```
Out[20] =  $u^{(1,0)}[\mathbf{x}, t]$ 
```

## ■ Stress $\sigma$

The following equation is designated as the **first magnetostrictive equation**. It is based on a linear model of magnetostriction. The classical stress-strain relationship is extended by a magnetostrictive relation, hence the following expression describes an association between mechanical and electromagnetic properties of the system::

```
In[21] :=  $\sigma[\mathbf{x}_-, t_-] := E R \epsilon[\mathbf{x}, t] - \theta B[t]$ 
```

With:

$\sigma$ : stress

$\epsilon$ : strain

ER: E-modulus of the magnetostrictive rod (under constant magnetic flux)

$\theta$ : "classical" magnetostrictive constant, unit  $\frac{\text{Pascal}}{\text{Tesla}}$

B: Magnetic flux density (SI unit Tesla) in the rod. Please note that B is assumed to be constant over the length of the rod. See Chapter xxx for details.

Verification:

```
In[22] :=  $\sigma[\mathbf{x}2, t2]$ 
```

```
Out[22] =  $-\theta B[t2] + E R u^{(1,0)}[\mathbf{x}2, t2]$ 
```

## ■ Differential equation for wave propagation

```
In[23] := newton2
```

```
Out[23] =  $E R u^{(2,0)}[\mathbf{x}, t] == \rho R u^{(0,2)}[\mathbf{x}, t]$ 
```

## ■ Solution u for displacement in magnetostrictive rod

The displacement is assumed to be function of position multiplied by function of time:

```
In[24] :=  $u[\mathbf{x}_-, t_-] := X[\mathbf{x}] * T[t]$ 
```

Function of time:

```
In[25] :=  $T[t_-] := e^{i w t}$ 
```

Verification:

```
In[26] := Simplify[newton2]
```

```
Out[26] =  $e^{i t w} (w^2 \rho R X[\mathbf{x}] + E R X''[\mathbf{x}]) == 0$ 
```

The term with the magnetic flux density disappeared because  $\partial_x B=0$  and according to Maxwell  $\text{div}(\mathbf{B})=0$

The function of position is assumed to be standing wave, the coefficients  $a_1$  and  $a_2$  can be complex (????????)

```
In[27] := X[x_] := a1 * Cos[K x] + a2 * Sin[K x]
```

The wave number K:

```
In[28] := K = w / cR
```

```
Out[28] = \frac{w}{cR}
```

With:

K: Wave number.

w: Circular frequency of oscillation.

cR: Sound velocity in magnetostrictive rod.

Verification:

Newton's second law (equation "newton2") will now be verified with the solutions defined for u:

```
In[29] := Simplify[newton2 /. cR -> \sqrt{ER / \rho R}]
```

```
Out[29] = True
```

## ■ Velocity v

Velocity is the first derivative of the displacement, with respect to time:

```
In[30] := v[x_, t_] = \partial_t u[x, t]
```

```
Out[30] = i e^{i t w} \left( a1 \cos\left[\frac{w x}{cR}\right] + a2 \sin\left[\frac{w x}{cR}\right] \right)
```

## ■ Electromagnetic relations (part 1)

### ■ Relation between Magnetic Flux Density B and properties of the solenoid

Based on Faraday's law the following equation is valid for an **ideal solenoid**:

```
In[31] := B[t_] := \frac{1}{i w N A} * U[t]
```

With:

- B: Magnetic flux density in the rod.  
 N: Number of turns of the solenoid  
 A: Cross-sectional area of the solenoid.

Please note that the magnetic flux density is related to the voltage, not to the current going through the solenoid!

Actually a constant magnetic field for biasing the magnetostrictive rod is excluded from this model. Therefore B and H mean just the alternating components.

### ■ Voltage drop function

The voltage drop at the solenoid has to be a sine-shaped function when the system is in steady state. Therefore the time function  $T[t]$  (former used for the displacement) is also used for the voltage. However, there may be a phase shift between the displacement and the voltage.

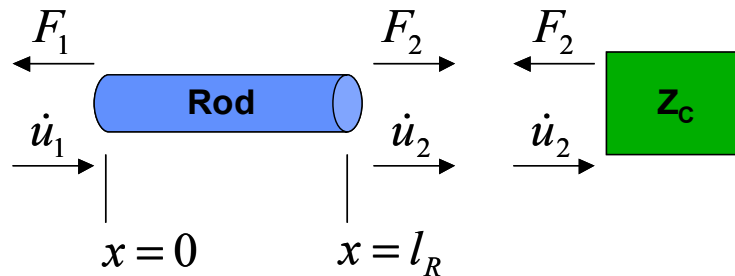
`In[32] := U[t_] := Ud * T[t]`

With:

Ud: Complex amplitude of the voltage function.

T:  $T = e^{i\omega t}$

### ■ Boundary conditions for amplitude coefficients $a_1$ and $a_2$



### ■ First boundary condition (determining coefficient $a_2$ )

No stress at free end of rod ( $x=0$ ) at all times:

`In[33] := BC1 = σ[0, t] == 0`

`Out[33] =  $\frac{a_2 e^{i\omega} E R \omega}{c R} + \frac{i e^{i\omega} U_d \theta}{A N \omega} == 0$`

Solve for coefficient  $a_2$  and create a rule:

```
In[34] := rulea2 = Flatten[Solve[BC1, a2]]
```

$$\text{Out}[34] = \left\{ a2 \rightarrow -\frac{i \, cR \, U d \, \theta}{A \, E R \, N \, w^2} \right\}$$

a2 is the amplitude of the sine function of the displacement, SI unit meter.

## ■ Second boundary condition (determining coefficient a1)

It is assumed that the sealant as well as the polymer is in contact with the rod only at the end of the rod (**keen simplification!**). Therefore an overall impedance ZC is defined as a second boundary condition:

$$\text{In}[35] := \text{BC2} = -ZC == Ac * \frac{\sigma[1Rod, t]}{v[1Rod, t]};$$

With:

ZC: Complex mechanical contact impedance of polymer and sealant, SI unit  $\frac{Ns}{m} = \frac{kg}{s}$

Ac: Contact area.

### Model simplifications:

$$\text{In}[36] := Ac = A;$$

Equals cross-sectional area of rod and solenoid in this model.

Derivation of a1:

$$\text{In}[37] := \text{BC2} = \text{Simplify}[\text{BC2}];$$

Solve for coefficient a1 and create a rule:

$$\text{In}[38] := \text{rulea1} = \text{Simplify}[\text{Flatten}[\text{Solve}[\text{BC2}, a1]]]$$

$$\text{Out}[38] = \left\{ a1 \rightarrow -\frac{cR \, U d \, \theta - i \, A \, a2 \, E R \, N \, w^2 \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] + a2 \, cR \, N \, w^2 \, ZC \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right]}{N \, w^2 \, (cR \, ZC \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] + i \, A \, E R \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right])} \right\}$$

Apply rule for coefficient a2 on coefficient a1:

$$\text{In}[39] := \text{rulea1} = \text{rulea1} /. \text{rulea2}$$

$$\text{Out}[39] = \left\{ a1 \rightarrow -\frac{cR \, U d \, \theta - cR \, U d \, \theta \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] - \frac{i \, cR^2 \, U d \, ZC \, \theta \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right]}{A \, E R}}{N \, w^2 \, (cR \, ZC \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] + i \, A \, E R \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right])} \right\}$$

$$\text{In}[40] := \text{rulea1} = \text{Simplify}[\text{rulea1}]$$

$$\text{Out}[40] = \left\{ a1 \rightarrow \frac{cR \, U d \, \theta \, (-A \, E R + A \, E R \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] + i \, cR \, ZC \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right])}{A \, E R \, N \, w^2 \, (cR \, ZC \, \text{Cos}\left[\frac{1Rod \, w}{cR}\right] + i \, A \, E R \, \text{Sin}\left[\frac{1Rod \, w}{cR}\right])} \right\}$$

a1 is the amplitude of the cosine function of the displacement, SI unit meter.

a1 in case of ZC=0:

```
In[41] := rulea1ZC0 = Simplify[rulea1 /. ZC -> 0]
```

```
Out[41] = {a1 ->  $\frac{i \, cR \, U d \, \theta \, \tan\left[\frac{lRod \, w}{2 \, cR}\right]}{A \, E R \, N \, w^2}$ }
```

Verification for the case that ZC=0 and w is first resonant frequency of rod:

```
In[42] := rulea1ZC0 /. w -> 2 \pi *  $\left(\frac{cR}{2 \, lRod}\right)$ 
```

```
Out[42] = {a1 -> ComplexInfinity}
```

Verification for the case that ZC=0 and w is second resonant frequency of rod:

```
In[43] := rulea1ZC0 /. w -> 2 \pi *  $\left(\frac{2 \, cR}{2 \, lRod}\right)$ 
```

```
Out[43] = {a1 -> 0}
```

Verification for limit values of ZC:

```
In[44] := Simplify[Limit[a1 /. rulea1, ZC -> \infty]]
```

```
Out[44] =  $\frac{i \, cR \, U d \, \theta \, \tan\left[\frac{lRod \, w}{cR}\right]}{A \, E R \, N \, w^2}$ 
```

```
In[45] := Simplify[Limit[a1 /. rulea1, ZC -> i * \infty]]
```

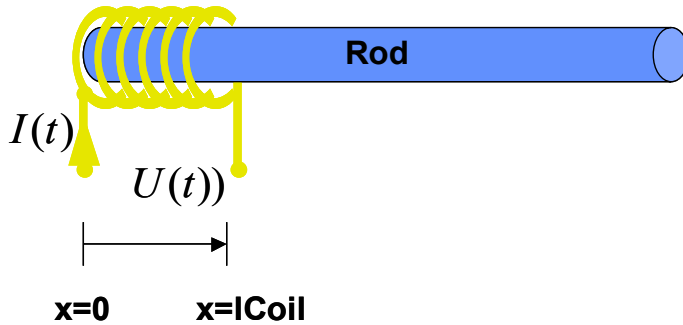
```
Out[45] =  $\frac{i \, cR \, U d \, \theta \, \tan\left[\frac{lRod \, w}{cR}\right]}{A \, E R \, N \, w^2}$ 
```

## ■ Absolute amplitude of oscillations:

This is the absolute amplitude of the oscillations:

```
In[46] := amp[a1_, a2_] :=  $\sqrt{Abs[a1]^2 + Abs[a2]^2}$ 
```

## ■ Further electromagnetic relations (part 2)



## ■ Magnetic field strength H

The following equation is designated as the **second magnetostrictive equation** of the linear model. The classical relationship between magnetic field strength and flux density ( $B = \mu H$ ) is extended by strain

$$In[47] := H[x_, t_] := \frac{1}{\mu R} * B[t] - \theta \epsilon[x, t]$$

With:

H: Magnetic field strength (SI unit  $\frac{\text{Ampere}}{\text{Meter}}$ ). The magnetic field strength

Verification:

$$In[48] := H[x, t]$$

$$Out[48] = -\frac{i e^{i t w} U d}{A N w \mu R} - e^{i t w} \theta \left( \frac{a2 w \cos\left[\frac{w x}{c R}\right]}{c R} - \frac{a1 w \sin\left[\frac{w x}{c R}\right]}{c R} \right)$$

Contrary to the flux density, the field strength varies in between the rod. Therefore H has to be integrated over the length of the coil. In the following consideration the solenoid is placed at the free end of the rod, with a yet unknown length  $l_{Coil}$ :

$$In[49] := Hq[t_] := \int_{x1}^{x1+l_{Coil}} \frac{H[x, t]}{l_{Coil}} dx$$

With:

Hq: Average magnetic flux density in the solenoid

$l_{Coil}$ : Length of the solenoid

$x1$ : Location of the coil

Verification:

In[50] := **Hq[t]**

$$\text{Out}[50] = \frac{1}{l_{\text{Coil1}}} \left( -\frac{i e^{i t w} l_{\text{Coil1}} U_d}{A N w \mu R} + a_1 e^{i t w} \theta \cos\left[\frac{w x_1}{c R}\right] - a_1 e^{i t w} \theta \cos\left[\frac{w (l_{\text{Coil1}} + x_1)}{c R}\right] + a_2 e^{i t w} \theta \sin\left[\frac{w x_1}{c R}\right] - a_2 e^{i t w} \theta \sin\left[\frac{w (l_{\text{Coil1}} + x_1)}{c R}\right] \right)$$

## ■ Current i

The current i is proportional to the (average) magnetic field strength, assuming an ideal solenoid:

$$\text{In}[51] := i[t_] := \frac{Hq[t] * l_{\text{Coil1}}}{N} / K1$$

With:

i: Current through solenoid

K1: A combined factor of geometry factor and field factor of the coil. For K=1 it becomes an equation for an ideal solenoid.

Verification:

In[52] := **i[t]**

$$\text{Out}[52] = \frac{1}{K1 N} \left( -\frac{i e^{i t w} l_{\text{Coil1}} U_d}{A N w \mu R} + a_1 e^{i t w} \theta \cos\left[\frac{w x_1}{c R}\right] - a_1 e^{i t w} \theta \cos\left[\frac{w (l_{\text{Coil1}} + x_1)}{c R}\right] + a_2 e^{i t w} \theta \sin\left[\frac{w x_1}{c R}\right] - a_2 e^{i t w} \theta \sin\left[\frac{w (l_{\text{Coil1}} + x_1)}{c R}\right] \right)$$

**This shows that the current is approximately proportional to the amplitude coefficient a1 (a2 is small near the resonance frequencies)!**

Including the displacement coefficients which were derived by using the boundary conditions:

In[53] := **ia[t\_] = FullSimplify[i[t] /. Flatten[{rulea1, rulea2}]]**

$$\begin{aligned} \text{Out}[53] = & -\left( e^{i t w} U_d \left( c R E R l_{\text{Coil1}} w Z C \cos\left[\frac{l_{\text{Rod}} w}{c R}\right] + \right. \right. \\ & i \left( A c R E R \theta^2 \mu R \cos\left[\frac{w (l_{\text{Rod}} - x_1)}{c R}\right] - A c R E R \theta^2 \mu R \cos\left[\frac{w x_1}{c R}\right] + \right. \\ & A c R E R \theta^2 \mu R \cos\left[\frac{w (l_{\text{Coil1}} + x_1)}{c R}\right] - A c R E R \theta^2 \mu R \cos\left[\frac{w (l_{\text{Coil1}} - l_{\text{Rod}} + x_1)}{c R}\right] + \\ & A E R^2 l_{\text{Coil1}} w \sin\left[\frac{l_{\text{Rod}} w}{c R}\right] + i c R^2 Z C \theta^2 \mu R \sin\left[\frac{w (l_{\text{Rod}} - x_1)}{c R}\right] + \\ & \left. \left. i c R^2 Z C \theta^2 \mu R \sin\left[\frac{w (l_{\text{Coil1}} - l_{\text{Rod}} + x_1)}{c R}\right] \right) \right) \right) / \\ & \left( A E R K1 N^2 w^2 \mu R \left( -i c R Z C \cos\left[\frac{l_{\text{Rod}} w}{c R}\right] + A E R \sin\left[\frac{l_{\text{Rod}} w}{c R}\right] \right) \right) \end{aligned}$$

With:

ia: Current through solenoid

Verification:



```
In[54] := (*FullSimplify[ia[t] /. {ZC -> 0, lCoil1 -> lRod}]) *
```

```
In[55] := (*FullSimplify[ia[t] /. {ZC -> 0}]) *
```

## ■ Electric Impedance Z

```
In[56] := Z := Simplify[U[t] / ia[t]]
```

With:

Z: Electric impedance of coil

Verification:

```
In[57] := Simplify[Z /. {ZC -> 0, lCoil1 -> lRod}]
```

```
Out[57] = 
$$\frac{i A E R K l N^2 w^2 \mu R \sin\left[\frac{lRod w}{cR}\right]}{-4 c R \theta^2 \mu R \cos\left[\frac{w x l}{cR}\right] \sin\left[\frac{lRod w}{2 cR}\right]^2 + E R lRod w \sin\left[\frac{lRod w}{cR}\right]}$$

```

## ■ Magnetostrictive power

Electric power which goes into the solenoid (complex - real part is active power, complex part is reactive power):

```
In[58] := P[t_] = U[t] * ia[t];
```

This is the **effective** (time independent) apparent power:

```
In[59] := Pa = 1 / 2 * Simplify[Abs[U[t]]] * Simplify[Abs[ia[t]], cond1];
```

## ■ Model verification

### ■ Unit Verification

### ■ Numeric calculations

## ■ Plotting

### ■ Amplitude at Constant Voltage (Artificial Case!)

```
In[79] := test1i = Abs[a1] /. valuerulea /. {w -> 2 π f, ZC -> 0, Ud -> 10} // N;
```

```
In[80] := test2i = Abs[a1] /. valuerulea /. {w -> 2 π f, ZC -> ZCexample1, Ud -> 10} // N;
```

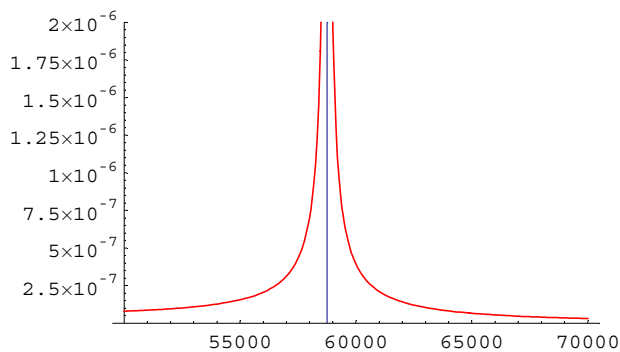
```
In[81]:= Plot[{test1i, test2i}, {f, 50000, 70000},
  PlotPoints → 150, PlotRange → {0, 2/1000000}, PlotStyle →
    {{Thickness[.004], RGBColor[1, 0, 0]}, {Thickness[.002], RGBColor[0, 1, 0]}},
  GridLines → {{58750, 117.5}, None}, Axes → {True, True}]
```

Plot::plnr : test2i is not a machine-size real number at f = 50000.00013422819`.

Plot::plnr : test2i is not a machine-size real number at f = 50130.685610436245`.

Plot::plnr : test2i is not a machine-size real number at f = 50273.20955659396`.

General::stop : Further output of Plot::plnr will be suppressed during this calculation.



Out[81]= - Graphics -

## ■ Impedance Plots

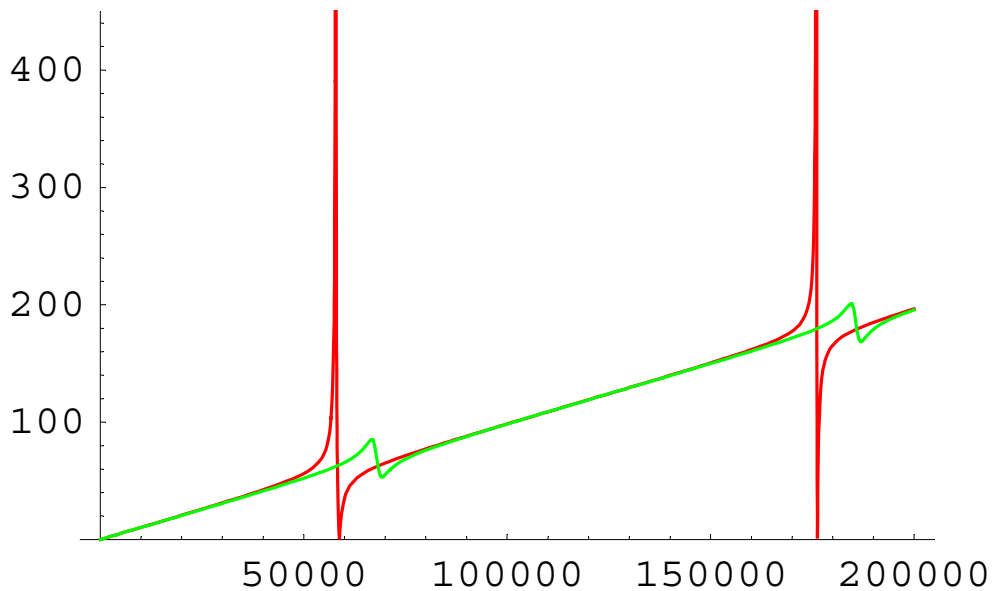
### ■ Plotting the electrical Impedance, once with long coil and once with short coil

### ■ Plotting the electrical Impedance, once without and once with load ZC:

```
In[85]:= test3 = Abs[Z /. lCoil1 → lRod /. values /. {ZC → 0, w → 2 π f}] // N;
```

```
In[86]:= test4 = Abs[Z /. lCoil1 → lRod /. values /. {ZC → 50 - i 300, w → 2 π f}] // N;
```

```
In[87]:= Plot[{test3, test4}, {f, 0, 200000}, PlotPoints → 150, PlotRange → {0, 450},
  AxesLabel → {"", ""}, TextStyle → {FontSize → 20}, PlotStyle →
  {{Thickness[.004], RGBColor[1, 0, 0]}, {Thickness[.004], RGBColor[0, 1, 0]}}
```



Out[87]= - Graphics -

## ■ Further electric relations (part 3)

### ■ Amplitude at HP Function Generator

The Voltage drop at the solenoid depends on the internal resistance and on the current:

```
In[88]:= Kirchhoff1 = U0[t] == i[t] (Ri + Z);
```

The internal voltage-source of the generator and the voltage at the solenoid have the same phase angle (because of the true ohmic internal resistance  $r_i$ ) and the same timefunction:

```
In[89]:= U0[t_] = U0d * T[t]
```

```
Out[89]= ei t w U0d
```

Solve for the voltage amplitude at the solenoid:

```
In[90]:= Kirchhoff1 = Simplify[Kirchhoff1 /. Flatten[{rulea1, rulea2}]];
```

```
In[91] := ruleUd = Flatten[Simplify[Solve[Kirchhoff1, Ud]]]
```

$$\begin{aligned} \text{Out}[91] = \left\{ \text{Ud} \rightarrow \left( A \text{ER} K1 N^2 U0d w^2 \mu R \left( -i cR ZC \cos\left[\frac{lRod w}{cR}\right] + A \text{ER} \sin\left[\frac{lRod w}{cR}\right] \right) \right) / \right. \\ \left( -cR \text{ER} w ZC (lCoil1 Ri + i A K1 N^2 w \mu R) \cos\left[\frac{lRod w}{cR}\right] - \right. \\ i A cR \text{ER} Ri \theta^2 \mu R \cos\left[\frac{w (lRod - x1)}{cR}\right] + i A cR \text{ER} Ri \theta^2 \mu R \cos\left[\frac{w x1}{cR}\right] - \\ i A cR \text{ER} Ri \theta^2 \mu R \cos\left[\frac{w (lCoil1 + x1)}{cR}\right] + i A cR \text{ER} Ri \theta^2 \mu R \cos\left[\frac{w (lCoil1 - lRod + x1)}{cR}\right] - \\ i A \text{ER}^2 lCoil1 Ri w \sin\left[\frac{lRod w}{cR}\right] + A^2 \text{ER}^2 K1 N^2 w^2 \mu R \sin\left[\frac{lRod w}{cR}\right] + \\ \left. cR^2 Ri ZC \theta^2 \mu R \sin\left[\frac{w (lRod - x1)}{cR}\right] + cR^2 Ri ZC \theta^2 \mu R \sin\left[\frac{w (lCoil1 - lRod + x1)}{cR}\right] \right) \left. \right\} \end{aligned}$$

Ud in case of constant lCoil:

```
In[92] := ruleUdlc = ruleUd /. values /. {lCoil1 → lRod} /. values;
```

```
In[93] := ruleUdlc2 = ruleUd /. values /. {lCoil1 → lRod / 2} /. values;
```

The following rule sets a1 and a2 for the HP function generator. Open is: lCoil1, Zp, w

```
In[94] := valueruleaHP = valuerulea /. ruleUd /. values;
```

In case of lcoil=lRod:

```
In[95] := valueruleaHPlc = Simplify[valuerulea /. ruleUd /. lCoil1 → lRod /. values];
```

In case of lcoil=lRod / 3:

```
In[96] := valueruleaHPlc3 = Simplify[valuerulea /. ruleUd /. lCoil1 → lRod / 3 /. values];
```

## ■ Plotting (part 2)

### ■ Amplitude and Phase

```
In[97] := test3 = Abs[a1 /. rulea1 /. values /. ZC → 0 /. w → 2 π f] // N
```

$$\begin{aligned} \text{Out}[97] = \frac{1}{f^2} (1.88477 \times 10^{-6} \\ \text{Abs}[\text{Ud} (-2.51327 \times 10^6 + 2.51327 \times 10^6 \cos[0.0000534739 f]) \csc[0.0000534739 f]]) \end{aligned}$$

```
In[98] := test1 = Abs[a1 /. valueruleaHPlc /. ZC → 0 /. w → 2 π f] // N;
```

```
In[99] := test1b = Abs[a1 /. valueruleaHPlc /. ZC → ZCexample1 /. w → 2 π f] // N;
```

```
General::spell :
Possible spelling error: new symbol name "test1b" is similar to existing symbols {test1, test1i}.
```

```
In[100] := test2 = Arg[a1 /. valueruleaHPlc /. ZC → 0 /. w → 2 π f] // N;
```

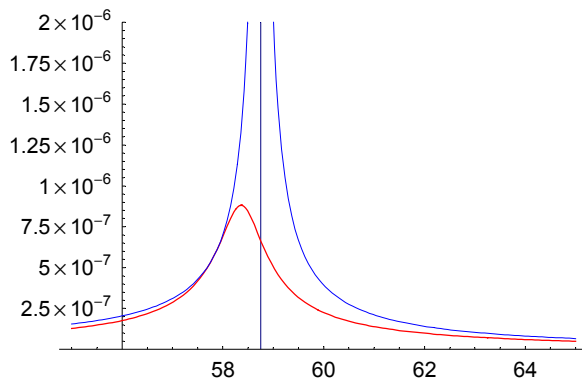
```
In[101]:= plla = Plot[{test1 /. f → fk*1000, test1b /. f → fk*1000,
  test1i /. f → fk*1000, test2i /. f → fk*1000}, {fk, 55, 65}, PlotPoints → 150,
  PlotRange → {0, 0.000002}, PlotStyle → {{Thickness[.003], RGBColor[1, 0, 0]},
  {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0, 1]},
  {Thickness[.002], RGBColor[1, 0, 1]}, {Thickness[.002], RGBColor[0.2, 0.8, 0]},
  {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0.6, 0.4]}}],
  TextStyle → {FontFamily → "Helvetica", FontSize → 10},
  GridLines → {{58.75, 117.5}, None}, Axes → {True, True}]
```

Plot::plnr : test1b /. f → fk 1000 is not a machine-size real number at fk = 55.000000067114094`.

Plot::plnr : test1b /. f → fk 1000 is not a machine-size real number at fk = 55.065342805218116`.

Plot::plnr : test1b /. f → fk 1000 is not a machine-size real number at fk = 55.13660477829698`.

General::stop : Further output of Plot::plnr will be suppressed during this calculation.



Out[101]= - Graphics -

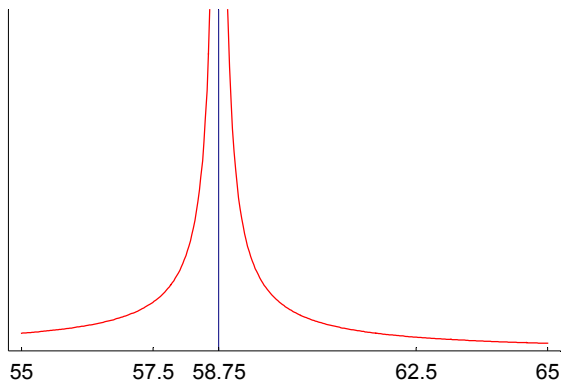
```
In[102]:= pl1b = Plot[{test1i /. f → fk*1000, test2i /. f → fk*1000},
  {fk, 55, 65}, PlotPoints → 350, PlotRange → {0, 0.000003},
  PlotStyle → {{Thickness[.003], RGBColor[1, 0, 0]},
    {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0, 1]},
    {Thickness[.002], RGBColor[1, 0, 1]}, {Thickness[.002], RGBColor[0.2, 0.8, 0]},
    {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0.6, 0.4]}}],
  TextStyle → {FontFamily → "Helvetica", FontSize → 10},
  GridLines → {{58.75, 117.5}, None}, Axes → {True, True},
  Ticks → {{55, 57.5, 58.75, 62.5, 65}, None}, AxesOrigin → {54.75, 0}]
```

Plot::plnr : test2i /. f → fk 1000 is not a machine-size real number at fk = 55.0000000286533`.

Plot::plnr : test2i /. f → fk 1000 is not a machine-size real number at fk = 55.027897071568766`.

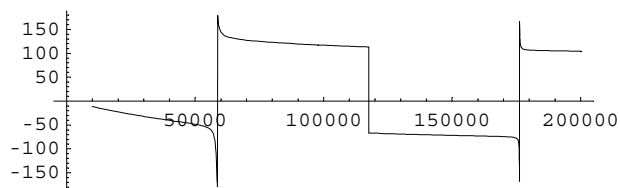
Plot::plnr : test2i /. f → fk 1000 is not a machine-size real number at fk = 55.05832123772564`.

General::stop : Further output of Plot::plnr will be suppressed during this calculation.



Out[102]= - Graphics -

```
In[103]:= pl2 = Plot[{test2 / (2 π) * 360}, {f, 10000, 200000},
  PlotPoints → 150, PlotRange → Automatic, AspectRatio → 1 / 3]
```



Out[103]= - Graphics -

## ■ Amplitude over Length-coordinate (Function X[x])

Plotting the imaginary part of the function X at a certain frequency (once without ZC, once with ZC≠0), over its coordinate x:

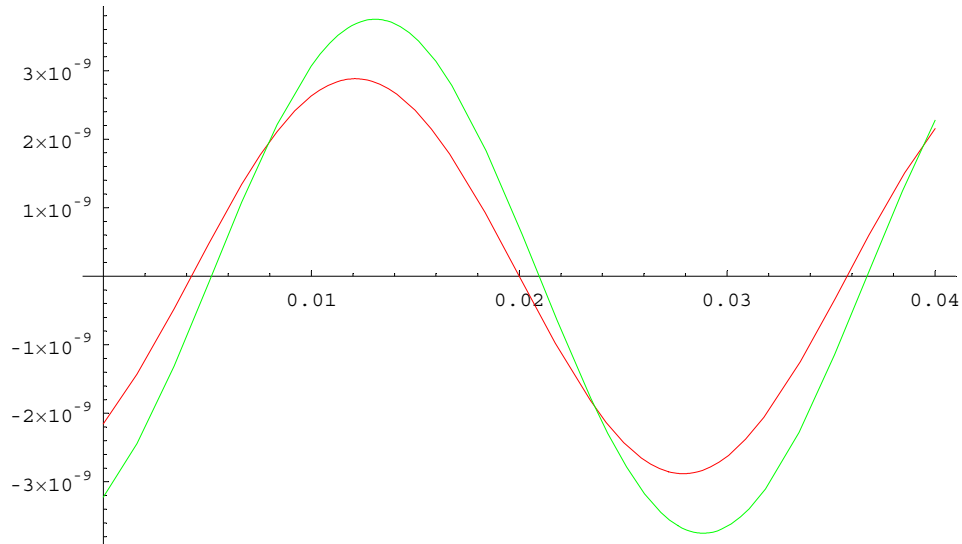
```
In[104]:= test1 =
  Im[X[x] /. Flatten[{rulea1, rulea2}] /. ruleUd1c /. values /. ZC → 0      /. w → 2 π f /.
  f → 149000] // N;
```

```

In[105]:= test2 =
  Im[X[x] /. Flatten[{rulea1, rulea2}] /. ruleUdlc /. values /. ZC → 0 + 250 i /. w → 2 π f /.
    f → 149000] // N;

In[106]:= Plot[{test1, test2}, {x, 0, lRod /. values},
  PlotStyle →
    {{Thickness[.001], RGBColor[1, 0, 0]}, {Thickness[.001], RGBColor[0, 1, 0]}}]

```



Out[106]= - Graphics -

Now plotting X over frequency:

```

In[107]:= test4 = Im[Xv /. w → 2 π f];

```

```
In[108]:= Plot3D[test4, {x, 0, lRod /. values}, {f, 10000, 210000}, PlotPoints -> 50]
```

```
Plot3D::plnc : test4 is neither a machine-size real number  
at {x, f}={0., 10000.} nor a list of a real number and a valid color directive.
```

```
Plot3D::plnc : test4 is neither a machine-size real number at  
{x, f}={0.000816327, 10000.} nor a list of a real number and a valid color directive.
```

```
Plot3D::plnc : test4 is neither a machine-size real number at  
{x, f}={0.00163265, 10000.} nor a list of a real number and a valid color directive.
```

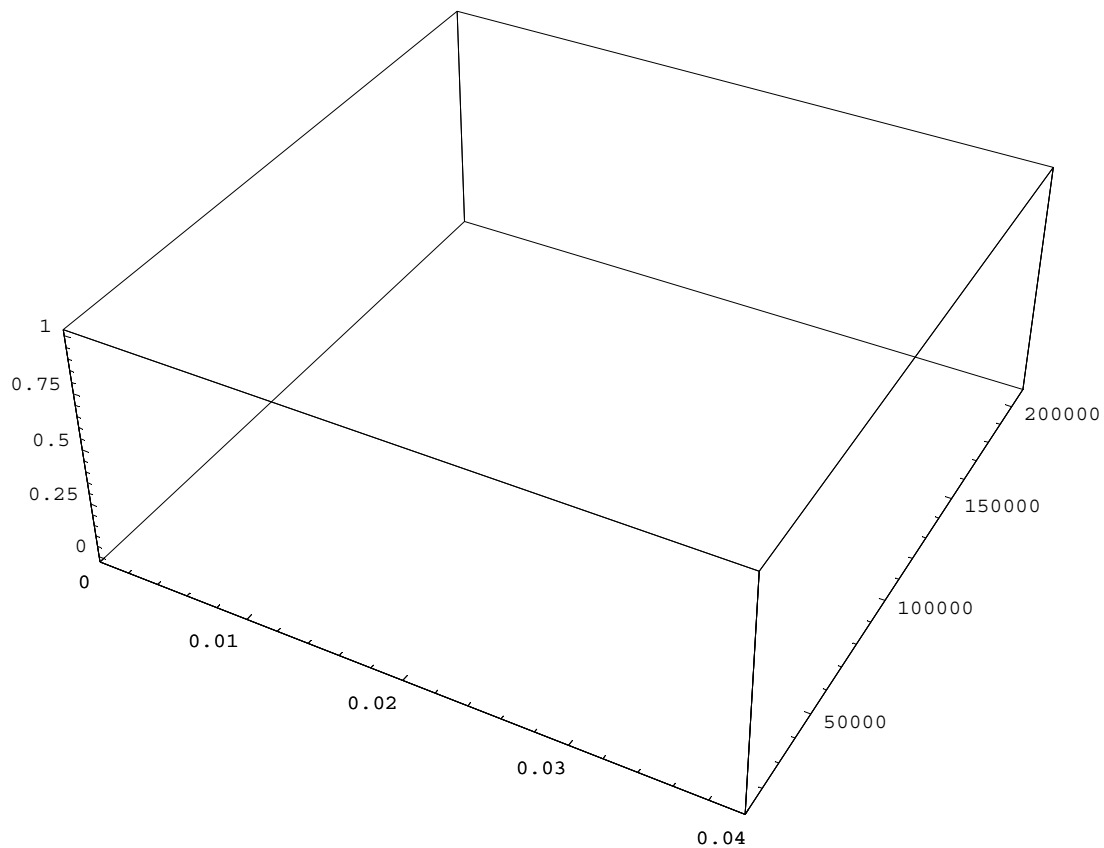
```
General::stop : Further output of Plot3D::plnc will be suppressed during this calculation.
```

```
Plot3D::gval : Function value Im[Xv] at grid point xi = 1, yi = 1 is not a real number.
```

```
Plot3D::gval : Function value Im[Xv] at grid point xi = 1, yi = 2 is not a real number.
```

```
Plot3D::gval : Function value Im[Xv] at grid point xi = 1, yi = 3 is not a real number.
```

```
General::stop : Further output of Plot3D::gval will be suppressed during this calculation.
```



```
Out[108]= - SurfaceGraphics -
```

## ■ Amplitude components (sin and cos)

ZC > 0



```
In[109]:= vas = a1 /. valueruleaHP1c
```

```
Out[109]= (5875 i π (-8000 π + 8000 π Cos[ $\frac{w}{117500}$ ] + 47 i ZC Sin[ $\frac{w}{117500}$ ])) /
```

$$\left( -47 i (-3125000 i w ZC + \pi^2 (1500000000000 + w^2 ZC)) \cos\left[\frac{w}{117500}\right] + \right.$$

$$\left. 2000 \pi (35250000000 i \pi + (-12500000 i w + 4 \pi^2 w^2 + 103546875 ZC) \sin\left[\frac{w}{117500}\right]) \right)$$

```
In[110]:= vas2 = a2 /. valueruleaHP1c
```

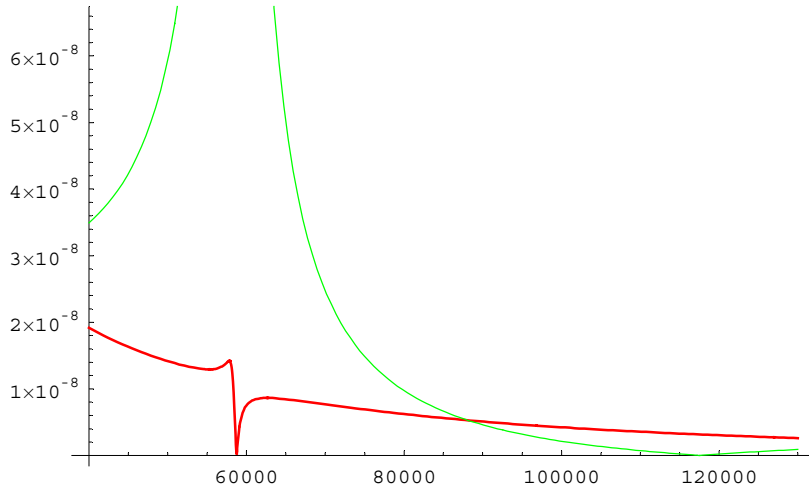
```
Out[110]= (5875 π (47 ZC Cos[ $\frac{w}{117500}$ ] + 8000 i π Sin[ $\frac{w}{117500}$ ])) /
```

$$\left( -47 i (-3125000 i w ZC + \pi^2 (1500000000000 + w^2 ZC)) \cos\left[\frac{w}{117500}\right] + \right.$$

$$\left. 2000 \pi (35250000000 i \pi + (-12500000 i w + 4 \pi^2 w^2 + 103546875 ZC) \sin\left[\frac{w}{117500}\right]) \right)$$

```
In[135]:= Plot[{Abs[vas2 /. ZC → 0 /. w → 2 π f], Abs[vas /. ZC → 0 /. w → 2 π f]},
```

$$\{f, 40000, 130000\}, \text{PlotPoints} \rightarrow 150, \text{PlotRange} \rightarrow \text{Automatic}, \text{PlotStyle} \rightarrow$$

$$\{\{\text{Thickness}[\text{.004}], \text{RGBColor}[1, 0, 0]\}, \{\text{Thickness}[\text{.002}], \text{RGBColor}[0, 1, 0]\}\}]$$


```
Out[135]= - Graphics -
```

Amplitude does not go to infinity, even under resonance conditions! This is related with the internal losses of the HP power supply. Setting the internal resistance  $R_i$  to zero would cause a constant voltage condition, with infinite current. See below and see above.

## ■ Absolute Amplitude Plots

### ■ Low mechanical impedance

Zero impedance (red):

```
In[112]:= vat1 = amp[a1, a2] /. valueruleaHP1c /. w → 2 π f * 1000 /. ZC → 0 // N;
```

Resistive impedance (green):

```
In[113]:= vat2 = amp[a1, a2] /. valueruleaHPlc /. w → 2 π f * 1000 /. ZC → 15 // N;
```

General::spell : Possible spelling error: new symbol name "vat2" is similar to existing symbols {val2, vas2}.

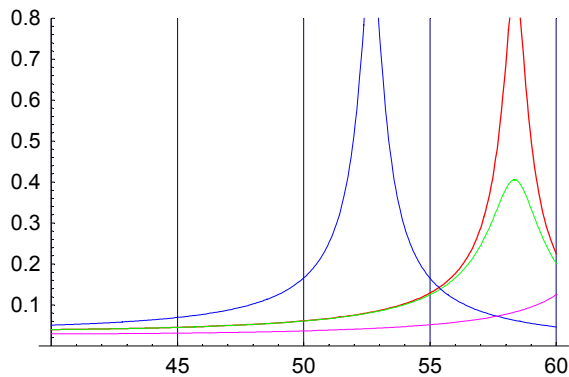
Resistive impedance + inertia term (blue):

```
In[114]:= vat3 = amp[a1, a2] /. valueruleaHPlc /. ZC → 0 + i w * 5 * 10-4 /. w → 2 π f * 1000 // N;
(*      ZC=15+i 30      *)
```

Resistive impedance + capacitive term (pink):

```
In[115]:= vat4 = amp[a1, a2] /. valueruleaHPlc /. ZC → -  $\frac{i}{2 * 10^{-8} w}$  /. w → 2 π f * 1000 // N;
(*      ZC=15-i 30      *)
```

```
In[116]:= p11 = Plot[{106 vat1, 106 vat2, 106 vat3, 106 vat4}, {f, 40, 60}, PlotPoints → 350,
  PlotRange → {0, 0.8}, PlotStyle → {{Thickness[.003], RGBColor[1, 0, 0]},
    {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0, 1]},
    {Thickness[.002], RGBColor[1, 0, 1]}, {Thickness[.002], RGBColor[0.2, 0.8, 0]},
    {Thickness[.002], RGBColor[0, 1, 0]}, {Thickness[.002], RGBColor[0, 0.6, 0.4]}}},
  TextStyle → {FontFamily → "Helvetica", FontSize → 10}, GridLines → {Automatic, None}]
```



Out[116]= - Graphics -

## ■ High mechanical impedance (limit values)

## ■ Receiving Coil

Note that the effect of a receiving coil is not mentioned in the equations above. Therefore they are only valid for a low power consumption at the receiving coil.

Index 2 for the receiving coil.

Faraday again:

```
In[123]:= U2[t] := i w N2 A B[t]
```

This shows that the voltage of the receiving coil is related to the voltage of exciting coil. This is the same equation like for a transformer. Therefore the magnetostrictive effect can only be measured because of the internal resistance of the voltage source.

```
In[124] := U2[t]
```

```
Out[124] = 
$$\frac{e^{i t \omega} N_2 U_d}{N}$$

```

An other method to detect the magnetostrictive effect is by measuring the current at the receiving coil:

```
In[125] := i2[t_] := 
$$\frac{Hq2[t] * lCoil2}{N2} * K2$$

```

With the average field strength:

```
In[126] := Hq2[t_] := 
$$\int_{x2}^{x2+lCoil2} \frac{H[x, t]}{lCoil2} dx$$

```

```
In[127] := i2a[t_] = Simplify[i2[t] /. rulea2 /. rulea1 /. ruleUd];
```

Now the power output of the receiving coil is calculated (complex - real part is active power, complex part is reactive power):

```
In[128] := P2[t] = U2[t] * i2a[t];
```

This is the **effective** (time independent) apparent power:

```
In[129] := P2a = 1 / 2 * Simplify[Abs[U2[t]], cond1] * Simplify[Abs[i2a[t]], cond1] /. ruleUd;
```

Further simplification, with dividing every complex variable into real and imaginary part (and into absolute part and argument, respectively):

```
In[130] := P2as = ComplexExpand[P2a /. ZC → ZCr + i ZCi, TargetFunctions -> {Re, Im}];
```

Note that "ComplexExpand" assumes all variables to be real. This means we define the amplitude U0d to be real.

With:

$ZC_R$

$ZC_i$

Further simplification:

```
In[131] := (*P2axs=Simplify[P2ax,cond1,Trig->False,TimeConstraint->50 ]*)
```

## ■ Zero Impedance -